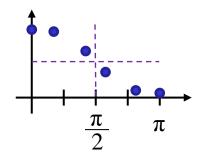
#### NYQUIST FILTERS

## Generation of Nyquist Filters

- Use remez(•) in matlab but you must constrain the frequency points and amplitudes in certain ways
  - The **frequency vector** values must mirror each other in pairs around  $\pi/2$ 
    - For example: [0 0.2 0.4 0.6 0.8 1]
      - [0 0.11 0.34 0.66 0.89 1]
  - The **amplitude vector** values must mirror each other in pairs around a magnitude of **0.50**
    - For example: [1 1 0 0] % low-pass [0 0.05 0.10 0.90 0.95 1] % high-pass
- Typically coefficients that should be zero will be close but not exactly zero when they are generated by remez(•)
  - Round these to make them exactly zero

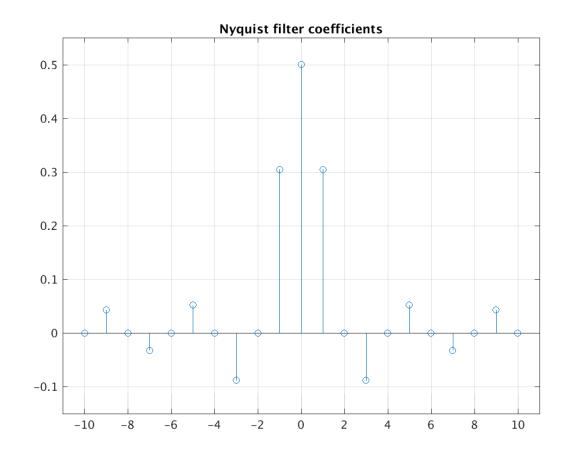


#### Nyquist Filter Example

 Copy and paste this program into a \*.m file and experiment yourself in matlab! % nyq.m 20 % 2015/03/04 Minor edits % 2018/03/12 Added scaling of figure 2 % Set these NumTaps = 21;PrintOn = 1; % Generate Nyquist filter coefficients coeffs = remez(NumTaps-1, [0 0.45 0.55 1], [1 0.95 0.05 0]); figure(1); clf; stem(-10:10, coeffs); axis ([-11 11 -0.15 0.55]); title('Nyquist filter coefficients'); grid on; if (PrintOn) print -dtiff 1.tiff; end figure(2); clf; freqz(coeffs); title('Filter frequency response plotted by freqz(); Note -6dB at \pi/2'); % select the top magnitude plot subplot(2,1,1); axis([0 1 -45 5]); % scale vertical axis more reasonably to see features hold on; plot(0.5, -6, 'ro'); if (PrintOn) print -dtiff 2.tiff; end % Generate white-noise flat-spectrum signal in = rand(1, 100000) - 0.5;figure(3); clf; psd(in); axis([0 1 -18 -4]); title('White-noise input signal to characterize filter; 100,000 samples'); if (PrintOn) print -dtiff 3.tiff; end % Pass the white-noise signal through the filter out = conv(coeffs, in); figure(4); clf; psd(out); title('Filter frequency response plotted by psd(); 100,000 samples; note -6dB (-17dB) at pi/2'); axis([0 1 -50 -5]); % scale vertical axis more reasonably to see features hold on; plot(0.5, -17, 'ro'); if (PrintOn) print -dtiff 4.tiff; end

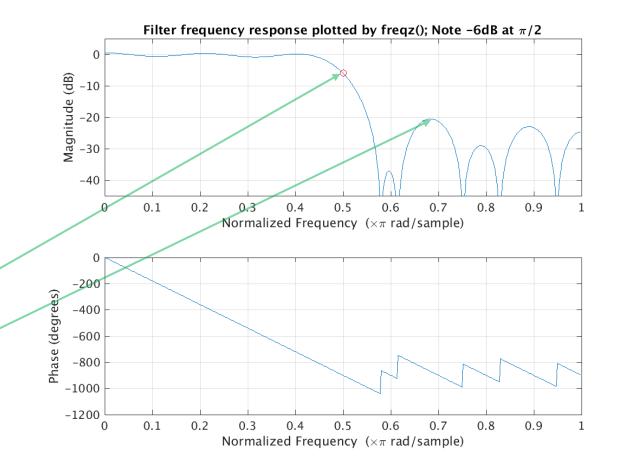
#### Nyquist Filter Coefficients Impulse Response

- 21-tap example
- It has significantly reduced hardware with almost half of its coeffs == zero
  - (N-1)/2 taps equal to zero for N = 4k+1
  - (N-3)/2 taps equal to zero for N = 4k+3



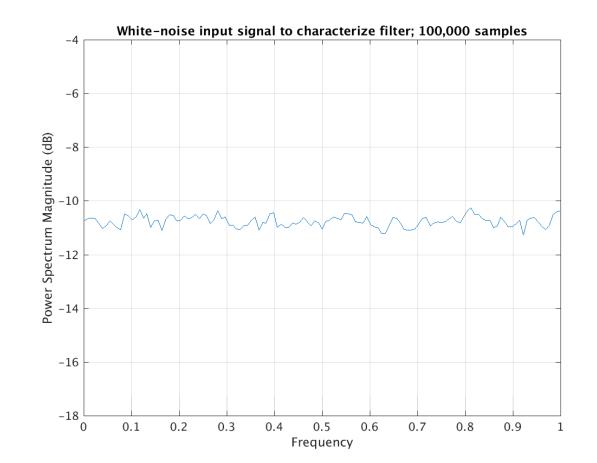
#### Filter Example

- The filter's frequency response plot made by freqz(•)
- Note these critical points to make a comparison later
  - $-6 \text{ dB at } \pi/2$ (1/2 magnitude)
  - -20 dB at
    0.68 π



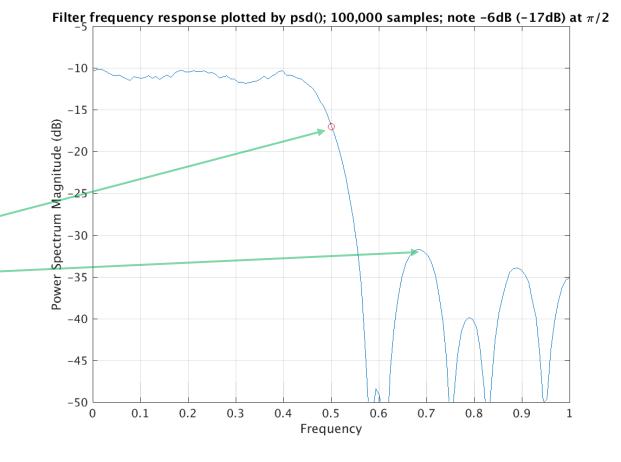
# The Second Less-Accurate Method to Measure Filter Response

- "White noise" random signals have a (nearly) flat spectrum
- This example contains 100,000 samples
  - More samples will make the spectrum flatter
- We will pass this signal through our filter and view the output spectrum to gauge the filter's frequency response



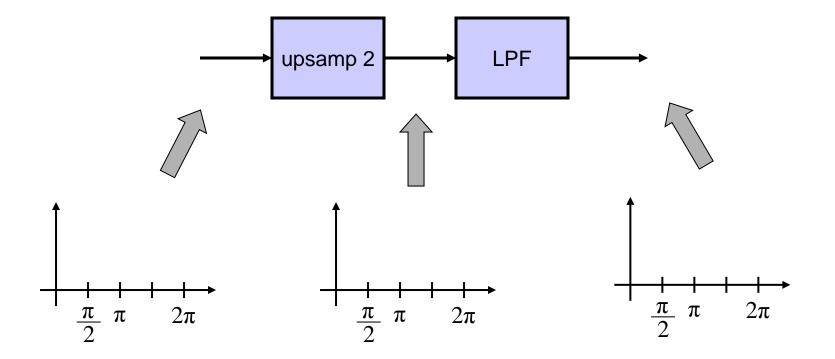
#### The Second Less-Accurate Method to Measure Filter Response

- This is the spectrum of the white-noise signal after being passed through the filter
- Note approximate values
  - -6 dB at  $\pi/2$
  - -20 dB at0.68  $\pi$
  - It matches freqz()!
- Recall that this method is best for actual bit-accurate HW designs



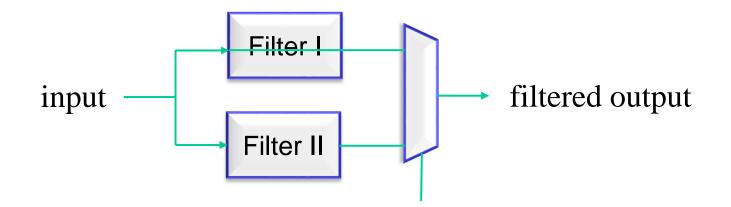
NYQUIST & UPSAMPLING FILTERS

- First recall standard approach
  - Low-pass filter has cutoff frequency at  $\pi/2$

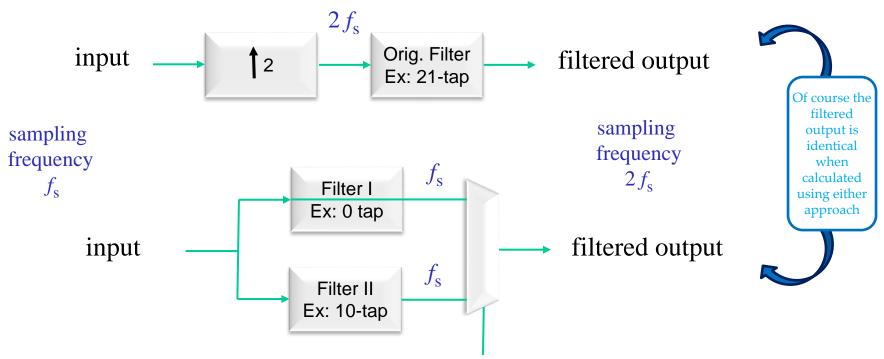


- Nyquist filters have almost half of their coefficients equal to zero
- Upsampled signals have every other sample equal to zero
- Lots of zeros ⇒ an opportunity!
- There are two alignments of data and filter coefficients
  - 1. The center tap of the filter aligns with a non-zero value in the upsampled data stream
    - The result is a trivial single multiply
    - With clever scaling, the multiplier can be reduced to a power-of-2 shift requiring no hardware whatsoever
  - 2. The other alignment
    - The result is a simplified filter with almost half the hardware because (*N*-1)/2 of the FIR multiplications are zero times zero
    - *N* delay registers are still needed however

• The Nyquist filter can then be implemented very efficiently by dividing the filter into two components that compute the two alignments and reconstructing the output with a 2:1 mux which interleaves samples taken from each filter every other cycle



- Another great benefit: the two filters are operating in the slower pre-upsampled sampling frequency domain
- Upsampling is performed in the mux



- Recall that for common static CMOS circuits,  $Power = C V^2 f$
- In summary, the optimized merged upsampler/filter using a Nyquist filter yields:
  - Approximately half the total hardware  $\rightarrow C' = C/2$
  - Filters operating at half the clock frequency  $\rightarrow f' = f/2$
  - Only a 2:1 mux operates at the faster upsampled clock frequency
  - *Power'* approximately 1/4 of the original power; probably a little less due to relaxed timing requirements of the simplified filter