EEC 118 Lecture #1: MOSFET Overview

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Jeff Parkhurst
Intel Corporation
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• Slides 13-17 Adapted from CSE477 VLSI Digital Circuits Lecture Slides by Vijay Narayanan and Mary Jane Irwin, Penn State University
Outline

• Administrative Details
• Survey of Digital IC Technology
• MOSFET Overview
Personnel

• **Prof. Raj Amirtharajah (Instructor)**
  Office: 3173 Kemper Hall
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  Please put EEC 118 in email subject line.
  Office Hours: Th 2 - 3 PM or by appointment.

• **Mackenzie Scott**
  Email: mrscott@ucdavis.edu
  Office Hours: (TBD)

• **Yixuan Zhai**
  Email: yxzhai@ucdavis.edu
  Office Hours: (TBD)

• **Labs**
  Tuesdays 5 PM – 8 PM 2157/2161 Kemper
  Wednesdays 5 PM – 8 PM 2157/2161 Kemper
Course Materials

• Textbook
  *Digital Integrated Circuits (2nd ed.)*
  by J. Rabaey, A. Chandrakasan, and B. Nikolic

• Suggested References
  *CMOS Digital Integrated Circuits (3rd ed.)* Kang and Leblebici
  *CMOS VLSI Design (3rd ed.)* Weste, Harris

• Handouts
  Labs, lab report cover sheets, slides, and lecture notes available on course web page in PDF format.

• Web Page
  Linked from MyUCDavis EEC 118 page and SmartSite
Grading

• Letter
  • A: 100 - 90%
  • B: 90 - 80%
  • C: 80 - 70%
  • D: 70 - 60%
  • F: below 60%

• Expect class average to be around B- / C+
• Curving will only help you
Weighting

- Labs 15%
- Design Project 15%
- Weekly Homework 5%
  Scale for each problem: 0 = poor effort, 1 = close, but fundamental problem, 2 = correct
- Quizzes 10%
  Four throughout the quarter (approx. every other week), lowest score dropped (April 13, April 22, May 20, June 1)
- Midterm 20%
  Wednesday, April 29
- Final 35%
  Tuesday, June 9, 8:00 - 10:00 AM
  Cumulative, but emphasizes material after midterm
Labs and CAD Software Usage

• Need to know/learn SPICE – Circuit Simulation
• Use same breadboard as EEC 180A
• No unsupervised lab hours!
  – TA or instructor must be present for your safety and security of the lab equipment
  – Extra lab hours will be added only in unusual circumstances
Education Demand for Circuit Design

• Industry needs circuit designers
  – Not just logic designers
    • Must understand operation at transistor level
  – Not just digital designers
    • Must understand analog effects
  – Not just analog designers
    • Must be able to comprehend Deep Sub-Micron (DSM) effects (<0.13um)

• Fundamental circuit knowledge critical
  – Similar techniques for bipolar transistors, NMOS (even relays and vacuum tubes!)
  – Must be able to exploit nanoscale devices in future
Education Demand for System Design

- Industry needs system designers
  - Need to understand system implications of your design
    - Power Delivery, Clock Loading – What do you need
  - Need to design from the system point of view
    - Communication protocol – how to effectively talk with other blocks
    - What should be added into your block to meet system design requirements (i.e. comprehend soft block methodology for optimization of area, interconnect, etc.)

You must operate at both levels!
Historical Background

Graph shows the growing complexity of designing integrated circuits
Memory, Processors and Graphics

- Used to be that memory and processors were the two main design drivers.
Memory, Processors and Graphics

- We now have graphics also driving integration

From ISPD 1999 Keynote Speech by Chris Malachowsky of NVIDIA
Hybrid to Monolithic Trend

- **We continue to integrate multiple functions on a single chip**
  - Mixture of Analog, Radio Frequency (RF), Digital
  - Graphics/Motherboard chipset an example of this

- **Cost and Performance driving market**
  - Higher performance achieved on chip than off chip
  - Lower cost due to a single die versus multi-chip design
  - Saves on packaging, total area by eliminating redundant functions

- **System-on-a-Chip (SOC) concept**
What are the issues facing the industry?

- Growth of transistors is exponential
- Growth of operating frequency is (was?) exponential
  - Reaching a limit due to power dissipation (see next generation Pentiums and Itaniums)
- Complexity continues to grow
  - Trend is toward multiple cores on one chip
  - Design teams cannot keep up with trend
- Power dissipation a concern
  - Power delivery, thermal issues, long term reliability
- Manufacturing providing us with lots of transistors
  - How do we use them effectively (besides large caches)?
Why worry about power? Power Dissipation

Lead microprocessors power continues to increase

Power delivery and dissipation will be prohibitive

Source: Borkar, De Intel®
Why worry about power? Chip Power Density

Power Density (W/cm²)

Year


10000

1000

100

10

1

8004 8008 8080 8085 286 386 486

atherajah/Parkhurst, EEC 118 Spring 2010 17

Source: Borkar, De Intel®
Chip Power Density Distribution

- Power density is not uniformly distributed across the chip
- Silicon not the best thermal conductor (isotopically pure diamond is)
- Max junction temperature is determined by hot-spots
  - Impact on packaging, cooling
Recent Battery Scaling and Future Trends

- Battery energy density increasing 8% per year, demand increasing 24% per year (Economist, January 6, 2005)
Why worry about power? Standby Power

- Drain leakage will increase as $V_T$ decreases to maintain noise margins and meet frequency demands, leading to excessive battery draining standby power consumption.

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<tbody>
<tr>
<td>Power supply $V_{dd}$ (V)</td>
<td>1.5</td>
<td>1.2</td>
<td>0.9</td>
<td>0.7</td>
<td>0.6</td>
</tr>
<tr>
<td>Threshold $V_T$ (V)</td>
<td>0.4</td>
<td>0.4</td>
<td>0.35</td>
<td>0.3</td>
<td>0.25</td>
</tr>
</tbody>
</table>

Source: Borkar, De Intel®
Emerging Microsensor Applications

Industrial Plants and Power Line Monitoring
(courtesy ABB)

Operating Room of the Future
(courtesy John Guttag)

Target Tracking & Detection
(Courtesy of ARL)

Location Awareness
(Courtesy of Mark Smith, HP)

Websign

NASA/JPL sensorwebs

Amirtharajah/Parkhurst, EEC 118 Spring 2010
Chip Design Styles

• **Field-Programmable Gate Array (FPGA)**
  – Regular structure. Not all transistors are usable.
  – Programmed via software (configurable wiring)

• **Gate Array**
  – Regular structure. Higher usage of transistors than FPGA
  – Two step manufacturing process.
    • Diffusion and poly initially. Design must be fairly stable
    • Metal layers fabricated once design is finalized

• **Cell based design**
  – All transistors used (may have spares to fill in area)
  – Each cell is fixed height so that they can be placed in rows

• **Full Custom**
  – Highest level of compactness and performance
  – Manually intensive. Not conducive to revision (ECO)
Logic Design Families

- **Static CMOS Logic**
  - Good power delay product (energy)
  - Good noise margin
  - Not as fast as dynamic

- **Dynamic Logic**
  - Very fast but inefficient in use of power
  - Domino, CPL, OPL

- **Pass Transistor Logic**
  - Poor noise margin
  - Sometimes static power dissipation
  - Less area than static CMOS
Design Parameters

• **Reliability** (Not dealt with when relating to layout)
  – Factors that dictate reliable operation of the circuit
    • Electromigration, thermal issues, hot electrons, noise margins

• **Performance** (Dealt with in this class)
  – Not just measured in clock speed. Power-Delay Product (PDP, equivalent to energy) is a better measure

• **Area** (Not dealt with when relating to layout)
  – Directly affects cost
Current State of the Art

• **Intel Core® @ 4 GHz (1 or 2 cores/chip going to 4+)**
  – 800 - 1066 MHz system bus
  – AGP 8x graphics (533 MHz bus)
  – Memory bus at 533 MHz (DDR)

• **Complex Designs demand resources**
  – Design teams resource limited due to logistics and cost
  – Cannot afford to miss issues due to cost of product recall
  – Emphasis on pre-silicon verification as opposed to post silicon testing
Modern Microprocessor
(> 100,000,000 transistors)
2003

Product Application:
AMD’s Opteron X86-64

- 8th generation processor (SledgeHammer) w/ 1MB L2 Cache
- Working on 1st (SOI) silicon; > 100 million transistors
- ~180mm² on 130nm technology with Cu metallization and low k
IBM POWER6

- Ultra-high frequency dual-core chip
  - 7-way superscalar, 2-way SMT core
  - 9 execution units
    - 2LS, 2FP, 2FX, 1BR, 1VMX, 1DFU
  - 790M transistors
  - Up to 64-core SMP systems
  - 2x4MB on-chip L2
  - 32MB On-chip L3 directory and controller
  - Two memory controllers on-chip
  - Recovery Unit
- Technology
  - CMOS 65nm lithography, SOI
- High-speed elastic bus interface at 2:1 freq
  - I/Os: 1953 signal, 5399 Power/Gnd

Reick et al., Hot Chips 19, 2007
Expectations

• **You should already know**
  – Solid State – (i.e. PN junctions, semiconductor physics, ..)

• **What we will cover**
  – MOS Transistors Fabrication and Equations
  – CMOS logic at the transistor level
  – Sequential logic
  – Memory
  – Arithmetic Circuits
  – Interconnect

• **Framework**
  – Course to use PowerPoint for the most part
  – Bring PowerPoint slides to class and write notes on them
MOS Transistor Types

- Rabaey Ch. 3 (Kang & Leblebici Ch. 3)
- Two transistor types (analogous to bipolar NPN, PNP)
  - NMOS: p-type substrate, n⁺ source/drain, electrons are charge carriers
  - PMOS: n-type substrate, p⁺ source/drain, holes are charge carriers
MOS Transistor Symbols

NMOS

PMOS
Note on MOS Transistor Symbols

- All symbols appear in literature
  - Symbols with arrows are conventional in analog papers
  - PMOS with a bubble on the gate is conventional in digital circuits papers
- Sometimes bulk terminal is ignored – implicitly connected to supply:

  ![NMOS Symbol](image)

  ![PMOS Symbol](image)

- Unlike physical bipolar devices, source and drain are usually symmetric
MOS Transistor Structure

- Important transistor physical characteristics
  - Channel length $L = L_{\text{gate}} - 2L_D$
  - Channel width $W$
  - Thickness of oxide $t_{\text{ox}}$
NMOS Transistor I-V Characteristics I

- I-V curve vaguely resembles bipolar transistor curves
  - Quantitatively very different
  - Turn-on voltage called **Threshold Voltage** $V_T$
• Drain current varies quadratically with gate-source voltage $V_{GS}$
MOS Transistor Operation: Cutoff

- **Simple case: \( V_D = V_S = V_B = 0 \)**
  - Operates as MOS capacitor (Cg = gate to channel)
  - Transistor in cutoff region

- **When \( V_{GS} < V_{T0} \), depletion region forms**
  - No carriers in channel to connect S and D (Cutoff)

![MOS Transistor Diagram](image-url)

\[ V_g < V_{T0} \]
\[ V_s = 0 \]
\[ V_d = 0 \]
\[ V_B = 0 \]
MOS Transistor Operation: Inversion

- When $V_{GS} > V_{T0}$, inversion layer forms
- Source and drain connected by conducting n-type layer (for NMOS)
  - Conducting p-type layer in PMOS
Threshold Voltage Components

• Four physical components of the threshold voltage

1. Work function difference between gate and channel (depends on metal or polysilicon gate): $\Phi_{GC}$

2. Gate voltage to invert surface potential: $-2\Phi_F$

3. Gate voltage to offset depletion region charge: $Q_B/C_{ox}$

4. Gate voltage to offset fixed charges in the gate oxide and oxide-channel interface: $Q_{ox}/C_{ox}$

$$C_{ox} = \frac{\mathcal{E}_{ox}}{t_{ox}} : \text{gate oxide capacitance per unit area}$$
Threshold Voltage Summary

• If \( V_{SB} = 0 \) (no substrate bias):

\[
V_{T0} = \Phi_{GC} - 2\phi_F - \frac{Q_{B0}}{C_{ox}} - \frac{Q_{ox}}{C_{ox}} \quad (K&L \ 3.20)
\]

• If \( V_{SB} \neq 0 \) (non-zero substrate bias)

\[
V_T = V_{T0} + \gamma \left( \sqrt{-2\phi_F + V_{SB}} - \sqrt{2\phi_F} \right) \quad (3.19)
\]

• Body effect (substrate-bias) coefficient:

\[
\gamma = \frac{\sqrt{2qN_A\varepsilon_{Si}}}{C_{ox}} \quad (K&L \ 3.24)
\]

• Threshold voltage increases as \( V_{SB} \) increases!
### Threshold Voltage (NMOS vs. PMOS)

<table>
<thead>
<tr>
<th></th>
<th>NMOS</th>
<th>PMOS</th>
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<tbody>
<tr>
<td>Substrate Fermi potential</td>
<td>$\phi_F &lt; 0$</td>
<td>$\phi_F &gt; 0$</td>
</tr>
<tr>
<td>Depletion charge density</td>
<td>$Q_B &lt; 0$</td>
<td>$Q_B &gt; 0$</td>
</tr>
<tr>
<td>Substrate bias coefficient</td>
<td>$\gamma &gt; 0$</td>
<td>$\gamma &lt; 0$</td>
</tr>
<tr>
<td>Substrate bias voltage</td>
<td>$V_{SB} &gt; 0$</td>
<td>$V_{SB} &lt; 0$</td>
</tr>
</tbody>
</table>
Body Effect

- **Body effect**: Source-bulk voltage $V_{SB}$ affects threshold voltage of transistor
  - Body normally connected to ground for NMOS, Vdd (Vcc) for PMOS
  - Raising source voltage increases $V_T$ of transistor
  - Implications on circuit design: series stacks of devices

If $V_x > 0$, $V_{SB} (A) > 0$, $V_T(A) > V_{TO}$
MOS Transistor Regions of Operation

- Three main regions of operation

- **Cutoff**: $V_{GS} < V_T$
  No inversion layer formed, drain and source are isolated by depleted channel. $I_{DS} \approx 0$

- **Linear (Triode, Ohmic)**: $V_{GS} > V_T$, $V_{DS} < V_{GS} - V_T$
  Inversion layer connects drain and source. Current is almost linear with $V_{DS}$ (like a resistor)

- **Saturation**: $V_{GS} > V_T$, $V_{DS} \geq V_{GS} - V_T$
  Channel is “pinched-off”. Current saturates (becomes independent of $V_{DS}$, to first order).
MOSFET Drain Current Overview

Saturation: \[ I_D = \frac{\mu C_{ox} W}{2} \frac{V}{L} \left( V_{GS} - V_T \right)^2 \left( 1 + \lambda V_{DS} \right) \]

Linear (Triode, Ohmic):

\[ I_D = \mu C_{ox} \frac{W}{L} \left( \left( V_{GS} - V_T \right)V_{DS} - \frac{V_{DS}^2}{2} \right) \]

Cutoff: \[ I_D \approx 0 \]

“Classical” MOSFET model, will discuss deep submicron modifications as necessary (Rabaey, Eqs. 3.25, 3.29)
A Fourth Region: Subthreshold

\[
I_D = I_S e^{\frac{V_{GS}}{n kT/q}} \left( 1 - e^{-\frac{V_{DS}}{kT/q}} \right)
\]

- Sometimes called “weak inversion” region
- When \( V_{GS} \) near \( V_T \), drain current has an exponential dependence on gate to source voltage
  - Similar to a bipolar device
- **Not typically used in digital circuits**
  - Sometimes used in very low power digital applications
  - Often used in low power analog circuits, e.g. quartz watches
Next Topic: MOSFET Details

- MOS Structure
  - Derivation of threshold voltage, drain current equations
- MOSFET Scaling
- MOSFET Capacitances